Advanced Systems Lab

Spring 2023

Lecture: Discrete Fourier transform, fast Fourier transforms

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Linear Transforms

Overview: Transforms and algorithms

Discrete Fourier transform

Fast Fourier transforms

After that:

- Optimized implementation and autotuning (FFTW)
- Automatic program synthesis (Spiral)

FFT References

Complexity: Bürgisser, Clausen, Shokrollahi, Algebraic Complexity Theory, Springer, 1997

History: Heideman, Johnson, Burrus: Gauss and the History of the Fast Fourier Transform, Arch. Hist. Sc. 34(3) 1985

FFTs:

- Cooley and Tukey, An algorithm for the machine calculation of complex Fourier series," Math. of Computation, vol. 19, pp. 297-301, 1965
- Nussbaumer, Fast Fourier Transform and Convolution Algorithms, 2nd ed., Springer, 1982
- van Loan, Computational Frameworks for the Fast Fourier Transform, SIAM, 1992
- Tolimieri, An, Lu, Algorithms for Discrete Fourier Transforms and Convolution, Springer, 2nd edition, 1997
- Franchetti, Püschel, Voronenko, Chellappa and Moura, Discrete Fourier Transform on Multicore, IEEE Signal Processing Magazine, special issue on ``Signal Processing on Platforms with Multiple Cores", Vol. 26, No. 6, pp. 90-102, 2009

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Linear Transforms

Very important class of functions: signal processing, communication, scientific computing, ...

Mathematically: Change of basis = Multiplication by a fixed matrix T

$$\begin{pmatrix} y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{pmatrix} = y = Tx$$

$$T = [t_{k,\ell}]_{0 \le k,\ell < n}$$

$$T = [t_{k,\ell}]_{0 \le k,\ell < n}$$
Input

Input

Equivalent definition: Summation form

$$y_k = \sum_{\ell=0}^{n-1} t_{k,\ell} x_{\ell}, \quad 0 \le k < n$$

Linear Transforms

x: input vector, y: output vector, T: fixed transform matrix Compute: y = Tx

Example: Discrete Fourier transform (DFT)

1. form (standard in signal processing):

given:
$$x_0,\ldots,x_{n-1}$$
 compute: $y_k=\sum_{\ell=0}^{n-1}e^{-2k\ell\pi i/n}x_\ell,\quad k=0,\ldots,n-1$ primitive nth root of 1
$$=\sum_{\ell=0}^{n-1}\omega_n^{k\ell}x_\ell,\quad k=0,\ldots,n-1,\quad \omega_n=e^{-2\pi i/n}$$

2. form (we will use):

given:
$$(x_0,\ldots,x_{n-1})^T$$
 compute: $y=\mathbf{DFT}_n\cdot x,\quad \mathbf{DFT}_n=[\omega_n^{k\ell}]_{0\leq k,\ell< n}$

How does the DFT₂ matrix look? Second row of DFT₄ matrix?

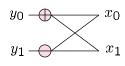
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Smallest Relevant Example: DFT, Size 2

Computation:
$$y = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

$$y = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} x$$
 or $y_0 = x_0 + x_1$ or $y_1 = x_0 - x_1$

As graph (direct acyclic graph or DAG):



called a butterfly



DFT, Size 4

$$DFT_4 = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i \end{bmatrix}$$

How many (complex) operations to compute the DFT_4 of a (complex) vector? $y = \mathrm{DFT_4} \cdot x$

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Transforms: Examples

A few dozen transforms are relevant

Some examples

$$\begin{array}{lll} \mathrm{DFT}_{n} &=& [e^{-2k\ell\pi i/n}]_{0 \leq k, \ell < n} \\ \mathrm{RDFT}_{n} &=& [r_{k\ell}]_{0 \leq k, \ell < n}, & r_{k\ell} = \begin{cases} \cos\frac{2\pi k\ell}{n}, & k \leq \lfloor\frac{n}{2}\rfloor \\ -\sin\frac{2\pi k\ell}{n}, & k > \lfloor\frac{n}{2}\rfloor \end{cases} & \textit{universal tool} \\ \mathrm{DHT} &=& \left[\cos(2k\ell\pi/n) + \sin(2k\ell\pi/n)\right]_{0 \leq k, \ell < n} \\ \mathrm{WHT}_{n} &=& \left[\frac{\mathrm{WHT}_{n/2}}{\mathrm{WHT}_{n/2}} - \frac{\mathrm{WHT}_{n/2}}{\mathrm{WHT}_{n/2}} \right], & \mathrm{WHT}_{2} = \mathrm{DFT}_{2} \\ \mathrm{IMDCT}_{n} &=& \left[\cos((2k+1)(2\ell+1+n)\pi/4n)\right]_{0 \leq k < 2n, 0 \leq \ell < n} & \textit{MPEG} \\ \mathrm{DCT-2}_{n} &=& \left[\cos(k(2\ell+1)\pi/2n)\right]_{0 \leq k, \ell < n} & \textit{JPEG} \\ \mathrm{DCT-3}_{n} &=& \mathrm{DCT-2}_{n}^{T} & (\mathrm{transpose}) \\ \mathrm{DCT-4}_{n} &=& \left[\cos((2k+1)(2\ell+1)\pi/4n)\right]_{0 \leq k, \ell < n} \end{array}$$

Transform Algorithms

An algorithm for y = Tx is given by a factorization

$$T = T_1 T_2 \cdots T_m$$

Namely, instead of y = Tx we can compute in steps

This reduces the op count only if:

- the T_i are sparse
- m is not too large

Example: Cooley-Tukey Fast Fourier Transform (FFT), size 4

$$\begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i \end{bmatrix} x = \begin{bmatrix} 1 & \cdot & 1 & \cdot \\ \cdot & 1 & \cdot & 1 \\ 1 & \cdot & -1 & \cdot \\ \cdot & 1 & \cdot & -1 \end{bmatrix} \begin{bmatrix} 1 & \cdot & \cdot & \cdot \\ \cdot & 1 & \cdot & \cdot \\ \cdot & \cdot & 1 & \cdot \\ \cdot & \cdot & \cdot & i \end{bmatrix} \begin{bmatrix} 1 & 1 & \cdot & \cdot \\ 1 & -1 & \cdot & \cdot \\ \cdot & \cdot & 1 & 1 \\ \cdot & \cdot & 1 & -1 \end{bmatrix} \begin{bmatrix} 1 & \cdot & \cdot & \cdot \\ \cdot & 1 & \cdot & \cdot \\ \cdot & 1 & \cdot & \cdot \\ \cdot & \cdot & 1 & -1 \end{bmatrix} x$$

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Cooley-Tukey FFT, n = 4

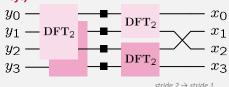
Fast Fourier transform (FFT)

$$\begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i \end{bmatrix} = \begin{bmatrix} 1 & \cdot & 1 & \cdot \\ \cdot & 1 & \cdot & 1 \\ 1 & \cdot & -1 & \cdot \\ \cdot & 1 & \cdot & -1 \end{bmatrix} \begin{bmatrix} 1 & \cdot & \cdot \\ \cdot & 1 & \cdot \\ \cdot & \cdot & 1 & \cdot \\ \cdot & \cdot & 1 & \cdot \\ \cdot & \cdot & \cdot & 1 \end{bmatrix} \begin{bmatrix} 1 & 1 & \cdot & \cdot \\ 1 & -1 & \cdot & \cdot \\ \cdot & \cdot & 1 & 1 \\ \cdot & \cdot & 1 & -1 \end{bmatrix} \begin{bmatrix} 1 & \cdot & \cdot & \cdot \\ \cdot & 1 & \cdot & \cdot \\ \cdot & \cdot & 1 & \cdot \\ \cdot & \cdot & \cdot & 1 \end{bmatrix}$$

Representation using matrix algebra

$$DFT_4 = (DFT_2 \otimes I_2) \operatorname{diag}(1, 1, 1, i) (I_2 \otimes DFT_2) L_2^4$$

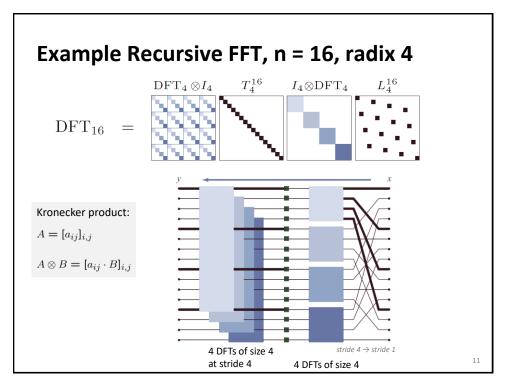
Data flow graph (right to left)

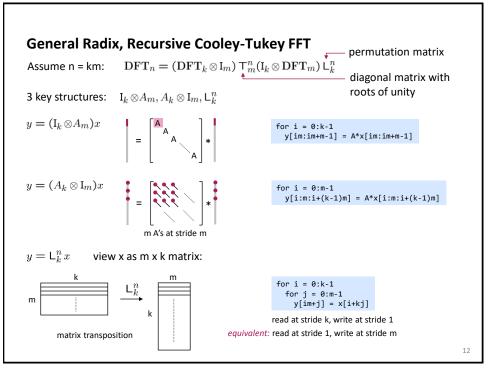


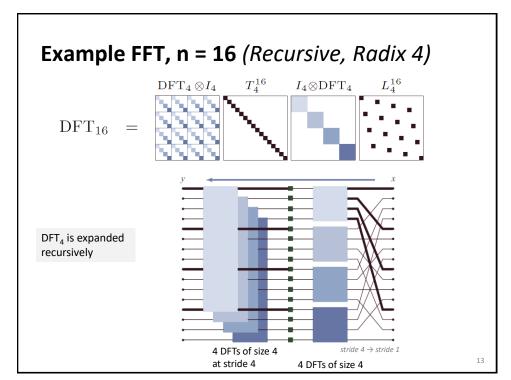
2 DFTs of size 2 at stride 2

2 DFTs of size 2

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Recursive Cooley-Tukey FFT

$$ext{DFT}_{km} = (ext{DFT}_k^* \otimes ext{I}_m) T_m^{km} (ext{I}_k \otimes ext{DFT}_m) L_k^{km}$$
 decimation-in-time

$$\mathrm{DFT}_{km} = L_m^{km} (\mathrm{I}_k \otimes \mathrm{DFT}_m) T_m^{km} (\mathrm{DFT}_k \otimes \mathrm{I}_m)$$
 decimation-in-frequency

For powers of two $n = 2^t$ sufficient together with base case

$$\mathbf{DFT}_2 = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

Cost:

- (complex adds, complex mults) = (n log₂(n), n log₂(n)/2) independent of recursion
- (real adds, real mults) ≤ (3n log₂(n), 2n log₂(n)) = 5n log₂(n) flops depends on recursion: best is at least radix-8

Recursive vs. Iterative FFT

Recursive, radix-k Cooley-Tukey FFT

$$DFT_{km} = (DFT_k \otimes I_m)T_m^{km}(I_k \otimes DFT_m)L_k^{km}$$

$$DFT_{km} = L_m^{km}(I_k \otimes DFT_m)T_m^{km}(DFT_k \otimes I_m)$$

Iterative, radix 2, decimation-in-time/decimation-in-frequency

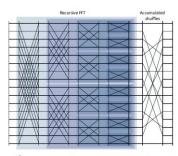
$$\mathbf{DFT}_{2^t} = \left(\prod_{j=1}^t (\mathbf{I}_{2^{j-1}} \otimes \mathbf{DFT}_2 \otimes \mathbf{I}_{2^{t-j}}) \cdot (\mathbf{I}_{2^{j-1}} \otimes T_{2^{t-j}}^{2^{t-j+1}})\right) \cdot R_{2^t}$$

$$\mathbf{DFT}_{2^t} = R_{2^t} \cdot \left(\prod_{j=1}^t (\mathbf{I}_{2^{t-j}} \otimes T_{2^{j-1}}^{2^j}) \cdot (\mathbf{I}_{2^{t-j}} \otimes \mathbf{DFT}_2 \otimes \mathbf{I}_{2^{j-1}}) \right)$$

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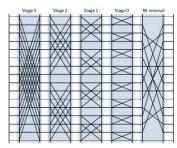
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 $\left(\mathrm{DFT}_2\otimes I_8\right)T_8^{16}\left(I_2\otimes\left((\mathrm{DFT}_2\otimes I_4)T_4^8\left(I_2\otimes\left((\mathrm{DFT}_2\otimes I_2)T_2^4(I_2\otimes\mathrm{DFT}_2)L_2^4\right)\right)L_2^8\right)\right)L_2^{16}$

Radix 2, iterative



 $((I_1 \otimes DFT_2 \otimes I_8)D_0^{16})((I_2 \otimes DFT_2 \otimes I_4)D_1^{16})((I_4 \otimes DFT_2 \otimes I_2)D_2^{16})((I_8 \otimes DFT_2 \otimes I_1)D_3^{16})R_2^{16}$

Recursive vs. Iterative

Iterative FFT computes in stages of butterflies = $log_2(n)$ passes through the data

Recursive FFT reduces passes through data = better locality

Same computation graph but different topological sorting

Rough analogy:

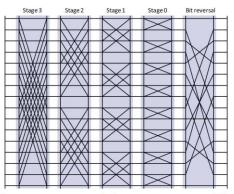
MMM	DFT
Triple loop	Iterative FFT
Blocked	Recursive FFT

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The FFT Is Very Malleable

Iterative FFT, Radix 2

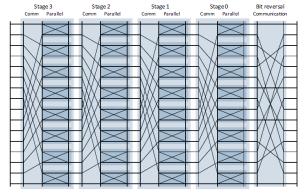


 $\Big(\big(I_1 \otimes \mathrm{DFT}_2 \otimes I_8\big)D_0^{16}\Big)\Big(\big(I_2 \otimes \mathrm{DFT}_2 \otimes I_4\big)D_1^{16}\Big)\Big(\big(I_4 \otimes \mathrm{DFT}_2 \otimes I_2\big)D_2^{16}\Big)\Big(\big(I_8 \otimes \mathrm{DFT}_2 \otimes I_1\big)D_3^{16}\Big)R_2^{16}$

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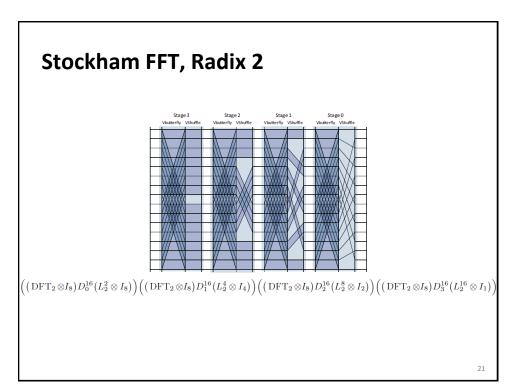
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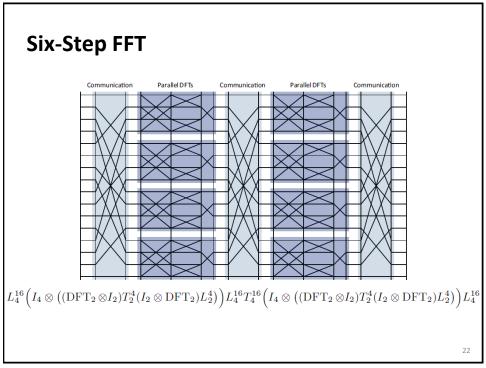
Pease FFT, Radix 2



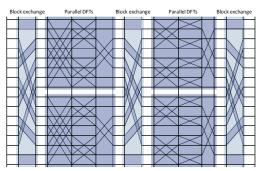
 $\Big(L_2^{16}\big(I_8 \otimes \mathrm{DFT}_2\big)D_0^{16}\Big)\Big(L_2^{16}\big(I_8 \otimes \mathrm{DFT}_2\big)D_1^{16}\Big)\Big(L_2^{16}\big(I_8 \otimes \mathrm{DFT}_2\big)D_2^{16}\Big)\Big(L_2^{16}\big(I_8 \otimes \mathrm{DFT}_2\big)D_3^{16}\Big)R_2^{16}$

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Multi-Core FFT



 $\left(L_4^8 \otimes I_2\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)L_2^4\right) \otimes I_2\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left(\mathrm{DFT}_2 \otimes I_2\right)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(L_2^8 \otimes I_2\right) T_4^{16} \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \mathrm{DFT}_2)\right)\right) R_2^8\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes I_2)\right)\right) R_2^8\right) R_2^8\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes I_2)\right)\right) R_2^8\right) R_2^8\right) \left(I_2 \otimes \left((\mathrm{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes I_2)\right) R_2^8\right) R_2^8\right) R_2^8\right) R_2^8\right) R_2^8\right) R_2^8\right) R_2^8\right) R_2^8$

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Transform Algorithms

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\mathbf{DFT}_n \to P_{k/2,2m}^\top \left( \mathbf{DFT}_{2m} \oplus \left( I_{k/2-1} \otimes_i C_{2m} \mathbf{rDFT}_{2m}(i/k) \right) \right) \left( \mathbf{RDFT}_k' \otimes I_m \right), \quad k \text{ even},
                   \begin{array}{l} \mathbf{RDFT}_n \\ \mathbf{RDFT}_n' \\ \mathbf{DHT}_n' \\ \mathbf{DHT}_n \\ \mathbf{DHT}_m' \\ \mathbf{DHT}_{2m} \\ \mathbf{DHT}_{2m} \\ \end{array} \\ + \left( P_{k/2,m}^T \otimes I_2 \right) \left( \begin{array}{c} \mathbf{RDFT}_{2m} \\ \mathbf{RDFT}_{2m}' \\ \mathbf{DHT}_{2m} \\ \mathbf{DHT}_{2m}' \\ \mathbf{DHT}_
 \begin{vmatrix} \mathbf{rDFT}_{2n}(u) \\ \mathbf{rDHT}_{2n}(u) \end{vmatrix} \rightarrow L_{m}^{2n} \left( I_{k} \otimes_{l} \begin{vmatrix} \mathbf{rDFT}_{2m}((i+u)/k) \\ \mathbf{rDHT}_{2m}((i+u)/k) \end{vmatrix} \right) \left( \begin{vmatrix} \mathbf{rDFT}_{2k}(u) \\ \mathbf{rDHT}_{2k}(u) \end{vmatrix} \otimes I_{m} \right), 
            \text{RDFT-3}_n \to (Q_{k/2,m}^\top \otimes I_2) \left(I_k \otimes_i \text{rDFT}_{2m}\right) (i+1/2)/k)) \left(\text{RDFT-3}_k \otimes I_m\right), \quad k \text{ even},
                  \mathbf{DCT} - \mathbf{2}_n \rightarrow P_{k/2,2m}^\top \left( \mathbf{DCT} - \mathbf{2}_{2m} K_2^{2m} \oplus \left( I_{k/2-1} \otimes N_{2m} \mathbf{RDFT} - \mathbf{3}_{2m}^\top \right) \right) B_n(L_{k/2}^{n/2} \otimes I_2) (I_m \otimes \mathbf{RDFT}_k') Q_{m/2,k},
                  \mathbf{DCT}\text{-}4_n \to Q_{k/2,2m}^\top \left(I_{k/2} \otimes N_{2m} \mathbf{RDFT}\text{-}3_{2m}^\top\right) B_n' (L_{k/2}^{n/2} \otimes I_2) (I_m \otimes \mathbf{RDFT}\text{-}3_k) Q_{m/2,k}.
                      \mathrm{DFT}_n \to (\mathrm{DFT}_k \otimes \mathrm{I}_m) \, \mathsf{T}_m^n (\mathrm{I}_k \otimes \mathrm{DFT}_m) \, \mathsf{L}_k^n, \quad n = km — Cooley-Tukey FFT
                        \mathrm{DFT}_n \to P_n(\mathrm{DFT}_k \otimes \mathrm{DFT}_m)Q_n, \quad n=km, \ \gcd(k,m)=1 Prime-factor FFT
                        \mathrm{DFT}_p \to R_p^T(\mathrm{I}_1 \oplus \mathrm{DFT}_{p-1})D_p(\mathrm{I}_1 \oplus \mathrm{DFT}_{p-1})R_p, \quad p \text{ prime}
            \operatorname{DCT-3}_n \to (\operatorname{I}_m \oplus \operatorname{J}_m) \operatorname{\mathsf{L}}_m^n (\operatorname{DCT-3}_m(1/4) \oplus \operatorname{DCT-3}_m(3/4))
                                                                                            \cdot (\mathsf{F}_2 \otimes \mathsf{I}_m) \begin{bmatrix} \mathsf{I}_m & 0 \oplus -\mathsf{J}_{m-1} \\ \frac{1}{\sqrt{2}} (\mathsf{I}_1 \oplus \mathsf{2} \, \mathsf{I}_m) \end{bmatrix}, \quad n = 2m
            DCT-4_n \rightarrow S_nDCT-2_n \operatorname{diag}_{0 \le k < n} (1/(2\cos((2k+1)\pi/4n)))
 \mathbf{IMDCT}_{2m} \ \rightarrow \ (\mathsf{J}_m \oplus \mathsf{I}_m \oplus \mathsf{I}_m \oplus \mathsf{J}_m) \bigg( \bigg( \begin{bmatrix} 1 \\ -1 \end{bmatrix} \otimes \mathsf{I}_m \bigg) \oplus \bigg( \begin{bmatrix} -1 \\ -1 \end{bmatrix} \otimes \mathsf{I}_m \bigg) \bigg) \ \mathsf{J}_{2m} \, \mathbf{DCT} - \mathbf{4}_{2m}
                 \mathbf{WHT}_{2^k} \rightarrow \prod_{i=1}^{r} (\mathbf{I}_{2^{k_1+\cdots+k_{i-1}}} \otimes \mathbf{WHT}_{2^{k_i}} \otimes \mathbf{I}_{2^{k_{i+1}+\cdots+k_t}}), \quad k = k_1 + \cdots + k_t
                      DFT_2 \rightarrow F_2
             DCT\text{-}\mathbf{2}_2 \ \rightarrow \ \text{diag}(1,1/\sqrt{2})\, \text{F}_2 
            DCT-4<sub>2</sub> \rightarrow J<sub>2</sub>R<sub>13\pi/8</sub>
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Complexity of the DFT

Measure: L_c , $2 \le c$

- Complex adds count 1
- Complex mults by a constant a with |a| < c counts 1
- L_2 is strictest, L_{∞} the loosest (and most natural)

Upper bounds:

■ $n = 2^k$: $L_2(DFT_n) \le 3/2 \text{ n } \log_2(n)$ (using Cooley-Tukey FFT)

■ General n: $L_2(DFT_n) \le 8 \text{ n } \log_2(n)$ (needs Bluestein FFT)

Lower bound:

- Theorem by Morgenstern: If $c < \infty$, then $L_c(DFT_n) \ge \frac{1}{2}$ n $\log_c(n)$
- Implies: in the measure L_{σ} , the DFT is $\Theta(n \log(n))$

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Lowest Known FFT Cost (Powers of 2)

A modified split-radix FFT with fewer arithmetic operations, *Johnson and Frigo, IEEE Trans. Signal Processing 55(1), pp. 111-119, 2007*

Number of flops $(n = 2^k)$:

$$\tfrac{34}{9} n \log_2(n) - \tfrac{124}{27} n - 2 \log_2(n) - \tfrac{2}{9} (-1)^{\log_2(n)} \log_2(n) + \tfrac{16}{27} (-1)^{\log_2(n)} + 8$$

History of FFTs

The advent of digital signal processing is often attributed to the FFT (Cooley-Tukey 1965)

History:

- Around 1805: FFT discovered by Gauss [1] (Fourier publishes the concept of Fourier analysis in 1807!)
- 1965: Rediscovered by Cooley-Tukey

[1]: Heideman, Johnson, Burrus: "Gauss and the History of the Fast Fourier Transform" Arch. Hist. Sc. 34(3) 1985

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Carl-Friedrich Gauss



1777 - 1855

Contender for the greatest mathematician of all times

Some contributions: Modular arithmetic, least square analysis, normal distribution, fundamental theorem of algebra, Gauss elimination, Gauss quadrature, Gauss-Seidel, non-Euclidean geometry, ...